

## APPENDIX E

LISTINGS OF THE ANALYTICAL AND SYNTHESIS PROGRAMS DEVELOPED TO  
STUDY AND DESIGN THE MULTIBAND ANTENNA

```

0010 PRINT "PROGRAM FOR CALCULATING VSWR ON GSRV-TYPE ANTENNAS"
0020 PRINT
0030 PRINT "UPDATED 12 JUNE 1985"
0040 PRINT
0050 PRINT "STORED ON DISC AS GSRV "
0060 PRINT "*****"
0070 DIM B(1000),B2(1000),B3(1000),Z(256,2),S3(100)
0080 PRINT "DO YOU WANT A FREQ. SWEEP (Y/N)?"
0090 INPUT I3
0100 IF I3="Y" THEN GOTO 0190
0110 PRINT "ENTER THE NUMBER OF FREQUENCIES TO BE TESTED"
0120 INPUT M
0130 PRINT "ENTER THE REQUIRED FREQUENCIES AND THEN 0 AS THE LAST VALUE"
0140 PRINT "AFTER EACH FREQUENCY TYPE NEW LINE"
0150 FOR I=1 TO (M+1)
0160   INPUT B(I)
0170   IF B(I)=0 THEN GOTO 0310
0180 NEXT I
0190 PRINT "ENTER LOWEST FREQ."
0200 INPUT A
0210 PRINT "ENTER HIGHEST FREQ."
0220 INPUT B
0230 PRINT "ENTER STEPSIZE"
0240 INPUT T
0250 LET M=1+(B-A)/T
0260 LET F=A
0270 FOR I=1 TO M
0280   LET B(I)=F
0290   LET F=F+T
0300 NEXT I
0310 OPEN FILE (3,3), "IMPEDANCES", 16
0320 FOR I=1 TO 250
0330   READ FILE (3,I-1), R,X
0340   LET Z(I,1)=R
0350   LET Z(I,2)=X
0360 NEXT I
0370 PRINT "*****BOTH L1 AND L2 ARE PHYSICAL LENGTHS *****"
0380 PRINT "ENTER L1"
0390 INPUT L1
0400 PRINT "ENTER L2"
0410 INPUT L2
0420 PRINT "ENTER VELOCITY FACTOR OF Z2"
0430 INPUT V
0440 LET L=L2/V
0450 PRINT "ENTER Z2"
0460 INPUT Z2
0470 PRINT "ENTER Z4"
0480 INPUT Z4
0490 LET B1=0
0500 FOR I=1 TO M
0510   LET F=B(I)
0520   IF F=0 THEN GOTO 0800
0530   LET K=(SYS(15)*L1*F)/300
0540   IF K<1 THEN GOTO 0790
0550   LET K1=(K-1)*10
0560   LET K2=INT(K1)
0570   LET R=Z(K2+1,1)
0580   LET X=Z(K2+1,2)
0590   LET R6=R
0600   LET X6=X
0610   LET R=Z(K2+2,1)
0620   LET X=Z(K2+2,2)
0630   LET R7=R

```

```

0640 LET I7=X
0650 LET K4=K1-K2
0660 LET R1=((R7-R6)*K4)+R6
0670 LET I1=((I7-I6)*K4)+I6
0680 LET B=(SYS(15)*F)/150
0690 LET T2=TAN(B*L)
0700 LET D=((1-(I1/Z2)*T2)^2+((R1/Z2)*T2)^2)
0710 LET R3=R1*(1+(T2)^2)/D
0720 LET N=(I1*(1-(T2)^2)+Z2*(1-(R1/Z2)^2-(I1/Z2)^2)*T2)
0730 LET I3=N/D
0740 LET P=SQR(((R3-Z4)^2+(I3)^2)/((R3+Z4)^2+(I3)^2))
0750 LET S=(1+P)/(1-P)
0760 LET B1=B1+1
0770 LET B2(B1)=F
0780 LET B3(B1)=S
0790 NEXT I
0800 PRINT "L1=";L1;"M"
0810 PRINT "L2(PHYSICAL)=";L2;"M"
0820 PRINT "Z2=";Z2;"OHMS"
0830 PRINT "Z4=";Z4;"OHMS"
0840 PRINT "VELOCITY FACTOR =" ;V
0850 PRINT
0860 PRINT
0870 FOR I=1 TO B1
0880 PRINT "F=";B2(I); "MHZ", "VSWR=";B3(I)
0890 PRINT
0900 NEXT I
0910 PRINT "ENTER HIGHEST VALUE OF VSWR TO BE PLOTTED"
0920 INPUT S2
0930 LET S1=S2-1
0940 PRINT "FREQ(MHZ)"; TAB(10); "I"; TAB(40); 1+S1/2; TAB(70); 1+S1; TAB(73); "VSWR"
0950 PRINT TAB(10); "I"-----[-----]
0960 FOR I=1 TO B1
0970 IF B3(I)>S2 THEN GOTO 1000
0980 PRINT B2(I); TAB(10); ":"; TAB(40); ":"; TAB(10+(B3(I)-1)*60/(S2-1)); ":"
0990 PRINT
1000 NEXT I
1010 PRINT " DO YOU WANT HARD COPY?"
1020 REM*****HARD COPY ONLY FOR TABLE NOT GRAPH
1030 INPUT S3
1040 IF S3="NO" THEN GOTO 1160
1050 OPEN FILE (4,2), "ALPT"
1060 PRINT FILE (4), "LISTING OF VSWR VS FREQ FOR ANTENNA AS SPECIFIED"
1070 PRINT FILE (4), "L1=";L1;"M"
1080 PRINT FILE (4), "L2(PHYSICAL)=";L2;"M"
1090 PRINT FILE (4), "Z2=";Z2;"OHMS"
1100 PRINT FILE (4), "Z4=";Z4;"OHMS"
1110 PRINT FILE (4), "VELOCITY FACTOR =" ;V
1120 FOR I=1 TO B1
1130 PRINT FILE (4), "F=";B2(I); "MHZ", "VSWR=";B3(I)
1140 NEXT I
1150 CLOSE FILE (4)
1160 CLOSE FILE (3)
1170 END

```

```

0010 PRINT "MULTIFREQUENCY ANTENNA ANALYSIS"
0020 PRINT "*****"
0030 PRINT
0040 PRINT "STORED ON DISK AS 1S6BKW "
0050 PRINT
0060 PRINT "BASED ON THE GSRV PRINCIPLE WITH IMPEDANCE DATA FROM KING/HARRISON"
0070 PRINT "ANTENNAS AND WAVES -A MODERN APPROACH"
0080 PRINT
0090 PRINT
0100 PRINT
0110 PRINT "ANTENNA LENGTH IS L1"
0120 PRINT "MATCHING STUB LENGTH IS L2"
0130 PRINT "STUB IMPEDANCE IS Z2"
0140 PRINT "TRANSMISSION LINE IMPEDANCE IS Z4"
0150 PRINT "ALL LENGTHS IN METRES"
0160 PRINT "FREQUENCY IN MHZ"
0170 PRINT
0180 PRINT
0190 PRINT
0200 DIM B(20),B2(20),B3(20),Z(256,2),S$(100)
0210 PRINT "ENTER THE NUMBER OF FREQUENCIES TO BE TESTED"
0220 INPUT M
0230 PRINT "WHAT IS THE MINIMUM NUMBER ACCEPTABLE FOR THE ANTENNA SYSTEM?"
0240 INPUT Q
0250 PRINT "WHAT IS THE MAXIMUM ACCEPTABLE VSWR?"
0260 INPUT W
0270 PRINT "ENTER THE REQUIRED FREQUENCIES AND THEN 0 AS THE LAST VALUE"
0280 PRINT "AFTER EACH FREQUENCY TYPE NEW LINE"
0290 FOR I=1 TO (M+1)
0300   INPUT B(I)
0310   IF B(I)=0 THEN GOTO 0330
0320 NEXT I
0330 OPEN FILE (3,3),"IMPEDANCES",16
0340 FOR I=1 TO 250
0350   READ FILE (3,I-1),R,X
0360   LET Z(I,1)=R
0370   LET Z(I,2)=X
0380 NEXT I
0390 PRINT "ENTER L1(MIN)"
0400 INPUT A
0410 PRINT "ENTER L1(MAX)"
0420 INPUT C
0430 PRINT "ENTER L1 STEP-SIZE"
0440 INPUT E
0450 PRINT "ENTER L2(MIN)"
0460 INPUT H
0470 PRINT "ENTER L2(MAX)"
0480 INPUT J
0490 PRINT "ENTER L2 STEP-SIZE"
0500 INPUT U
0510 FOR L1=A TO C STEP E
0520   FOR L2=H TO J STEP U
0530     FOR Z2=300 TO 600 STEP 50
0540       FOR Z4=50 TO 72 STEP 22
0550         LET B1=0
0560         FOR I=1 TO M
0570           LET F=B(I)
0580           IF F=0 THEN GOTO 0670
0590           LET K=(SYS(15)*L1*F)/300
0600           IF K<1 THEN GOTO 0660
0610           LET K1=(K-1)*10
0620           LET K2=INT(K1)
0630           LET R=Z(K2+1,1)

```

```

0640      LET I=Z(K2+1,2)
0650      LET R6=R
0660      LET I6=I
0670      LET R=Z(K2+2,1)
0680      LET I=Z(K2+2,2)
0690      LET R7=R
0700      LET I7=I
0710      LET K4=K1-K2
0720      LET R1=((R7-R6)*K4)+R6
0730      LET I1=((I7-I6)*K4)+I6
0740      LET B=(SYS(15)*F)/150
0750      LET T2=TAN(B*L2)
0760      LET D=((1-(I1/Z2)*T2)^2+((R1/Z2)*T2)^2)
0770      LET R3=R1*(1+(T2)^2)/D
0780      LET N=(I1*(1-(T2)^2)+Z2*(1-(R1/Z2)^2-(I1/Z2)^2)*T2)
0790      LET I3=N/D
0800      LET P=SQR(((R3-Z4)^2+(I3)^2)/((R3+Z4)^2+(I3)^2))
0810      LET S=(1+P)/(1-P)
0820      IF S>W THEN GOTO 0860
0830      LET B1=B1+1
0840      LET B2(B1)=F
0850      LET B3(B1)=S
0860      NEXT I
0870      IF B1<Q THEN GOTO 1100
0880      PRINT "THE FOLLOWING VALUES MEET THE SPECIFICATION VSWR<=",W
0890      PRINT "L1=";L1;"M"
0900      PRINT "L2=";L2;"M"
0910      PRINT "Z2=";Z2;"OHMS"
0920      PRINT "Z4=";Z4;"OHMS"
0930      FOR I=1 TO B1
0940          PRINT "F=";B2(I);"MHZ","VSWR=";B3(I)
0950          PRINT
0960      NEXT I
0970      PRINT " DO YOU WANT HARD COPY?"
0980      INPUT S$
0990      IF S$="NO" THEN GOTO 1100
1000      OPEN FILE (4,2),"#LPT"
1010      PRINT FILE (4),"THESE VALUES MEET THE SPECIFICATION VSWR<=";W
1020      PRINT FILE (4),"L1=";L1;"M"
1030      PRINT FILE (4),"L2=";L2;"M"
1040      PRINT FILE (4),"Z2=";Z2;"OHMS"
1050      PRINT FILE (4),"Z4=";Z4;"OHMS"
1060      FOR I=1 TO B1
1070          PRINT FILE (4),"F=";B2(I);"MHZ","VSWR=";B3(I)
1080      NEXT I
1090      CLOSE FILE (4)
1100      NEXT Z4
1110      NEXT Z2
1120      NEXT I,2
1130      NEXT L1
1140      CLOSE FILE (3)
1150      PRINT "MC LENGTHS WITHIN RANGE MEET THE SPECIFICATION"
1160      END

```



```

0010 PRINT "                                FEED3"
0020 PRINT "::::::::::::::::::::::::::::::::::::::::::::::::::::::::::"
0030 PRINT "FEED3 TO FIND Z1 GIVEN Z3"
0040 PRINT
0050 PRINT
0060 DIM B(100),B2(100),B3(100),B4(100),S9(100)
0070 DIM B1(100),B5(100),B6(100)
0080 PRINT "ENTER THE NUMBER OF FREQUENCIES AND IMPEDANCES TO BE TESTED"
0090 INPUT M
0100 PRINT "ENTER THE REQUIRED FREQUENCIES AND THEN 0 AS THE LAST VALUE"
0110 PRINT "AFTER EACH VALUE TYPE NEW LINE"
0120 FOR I=1 TO (M+1)
0130   INPUT B(I)
0140   IF B(I)=0 THEN GOTO 0160
0150 NEXT I
0160 PRINT "ENTER VALUES OF Z3 AND THEN 0 AS LAST VALUE"
0170 FOR I=1 TO (M+1)
0180   INPUT B1(I)
0190   IF B1(I)=0 THEN GOTO 0210
0200 NEXT I
0210 PRINT "ENTER VALUES OF A3 AND THEN 360 AS LAST VALUE"
0220 FOR I=1 TO (M+1)
0230   INPUT B5(I)
0240   IF B5(I)=360 THEN GOTO 0260
0250 NEXT I
0260 PRINT "ENTER PHYSICAL LENGTH OF ANTENNA L1"
0270 INPUT L1
0280 PRINT "ENTER PHYSICAL LENGTH L OF STUB"
0290 INPUT L
0300 PRINT "ENTER VELOCITY FACTOR V OF STUB"
0310 INPUT V
0320 LET L2=L/V
0330 PRINT "ENTER CHARACTERISTIC IMPEDANCE OF STUB = Z2"
0340 INPUT Z2
0350 PRINT "WHAT IS ANTENNA HEIGHT?"
0360 INPUT H5
0370 PRINT "DO YOU WANT ABSOLUTE OR NORMALISED VALUES OF Z1 W.R.T. Z2"
0380 PRINT "TYPE A OR N"
0390 INPUT A3
0400 IF A3="A" THEN GOTO 0450
0410 LET N1=Z2
0420 LET P5=1
0430 REM N1 AND P5 ARE MERELY MARKERS TO INDICATE ABS. OR NORM. VALUES
0440 GOTO 0460
0450 LET N1=1
0460 LET B1=0
0470 FOR I=1 TO M
0480   LET F=B(I)
0490   LET A3=B5(I)
0500   LET A3=(SYS(15)*A3)/150
0510   LET Z3=B1(I)
0520   LET R3=Z3*COS(A3)
0530   LET I3=Z3*SIN(A3)
0540   LET B=(SYS(15)*F)/150
0550   LET T2=TAN(B*L2)
0560   LET D=((Z2*I3*T2)^2+(R3*T2)^2)
0570   LET R1=((R3*Z2)*(Z2*(I3*T2))-(I3*Z2-(Z2^2*T2)*(R3*T2))/D
0580   LET R1=R1/N1
0590   LET I1=((R3*Z2)*(R3*T2)+(I3*Z2-(Z2^2*T2)*(Z2*I3*T2))/D
0600   LET I1=I1/N1
0610   LET K9=SYS(15)*L1*F/300
0620   LET B1=B1+1
0630   LET B2(B1)=F

```

```

0640 LET B3(B1)=R1
0650 LET B4(B1)=I1
0660 LET B6(B1)=INT(E9*100)/100
0670 NEXT I
0680 PRINT "L1=";L1;"M (PHYSICAL)"
0690 PRINT "L2=";L2;"M (ELECTRICAL)"
0700 PRINT "Z2=";Z2;"OHMS"
0710 PRINT "VELOCITY FACTOR =" ;V
0720 PRINT "ANTENNA HEIGHT=";H5;"M"
0730 PRINT
0740 PRINT
0750 PRINT
0760 PRINT
0770 PRINT "THESE VALUES ARE NORMALISED W.R.T. Z2"
0780 PRINT "*****"
0790 PRINT
0800 PRINT
0810 PRINT
0820 FOR I=1 TO B1
0830 PRINT "F=";B2(I); "MHZ", "R1=";B3(I); "OHMS", "I1=";B4(I); "OHMS", "XOH=";B6(I)
0840 PRINT
0850 NEXT I
0860 PRINT " DO YOU WANT HARD COPY?"
0870 INPUT S3
0880 IF S4="NO" THEN GOTO 1070
0890 OPEN FILE (4,2), "LPT"
0900 IF P5=1 THEN GOTO 0930
0910 PRINT FILE (4), "LISTING OF R1 AND I1 VS FREQ FOR ANTENNA AS SPECIFIED"
0920 GOTO 0940
0930 PRINT FILE (4), "NORMALISED VALUES OF R1 AND I1 W.R.T. Z2 VS FREQ."
0940 PRINT FILE (4), "L1=";L1;"M (PHYSICAL)"
0950 PRINT FILE (4), "L2=";L2;"M (ELECTRICAL)"
0960 PRINT FILE (4), "Z2=";Z2;"OHMS"
0970 PRINT FILE (4), "VELOCITY FACTOR =" ;V
0980 PRINT FILE (4), "ANTENNA HEIGHT=";H5;"M"
0990 PRINT FILE (4)
1000 PRINT FILE (4)
1010 PRINT FILE (4)
1020 PRINT FILE (4)
1030 FOR I=1 TO B1
1040 PRINT FILE (4), "F=";B2(I); "MHZ", "R1=";B3(I); "OHMS", "I1=";B4(I); "OHMS", "XO
1050 NEXT I
1060 CLOSE FILE (4)
1070 END

```

LISTING OF  $L_1$ ,  $L_2$  AND  $Z_2$  VALUES WHICH YIELDED VSWR  $\leq 2:1$  ON 7; 14; 18;  
24 AND 29 MHz BANDS

| $L_1$<br>(m) | $L_2$<br>(m) | $Z_2$<br>( $\Omega$ )   |
|--------------|--------------|-------------------------|
| 27,7         | 13,7         | 400                     |
| 27,9         | 13,6         | 400                     |
| 28,0         | 13,6         | 400; 375;               |
| 28,1         | 13,5         | 400;                    |
| 28,2         | 13,5         | 400; 375;               |
| 28,3         | 13,5         | 400; 375; 350           |
| 28,3         | 13,4         | 400                     |
| 28,4         | 13,5         | 400; 375;               |
| 28,4         | 13,4         | 400; 375; 350;          |
| 28,5         | 13,4         | 400; 375; 350; 325;     |
| 28,5         | 13,3         | 375;                    |
| 28,6         | 13,3         | 400; 375; 350           |
| 28,6         | 13,4         | 400; 375; 350;          |
| 28,6         | 13,2         | 400                     |
| 28,7         | 13,4         | 400; 375; 350           |
| 28,7         | 13,3         | 400; 375; 350; 325      |
| 28,7         | 13,2         | 400; 375; 350           |
| 28,8         | 13,3         | 400; 375; 350; 325      |
| 28,8         | 13,2         | 400; 375; 350; 325      |
| 28,8         | 13,1         | 400; 375                |
| 28,9         | 13,3         | 375; 250; 300           |
| 28,9         | 13,2         | 400; 375; 350; 325; 300 |
| 28,9         | 13,1         | 400; 375; 350; 325      |
| 28,9         | 13,0         | 375; 350                |
| 29,0         | 13,3         | 350; 325                |
| 29,0         | 13,2         | 375; 350; 325; 300      |
| 29,0         | 13,1         | 400; 375; 350; 325; 300 |
| 29,0         | 13,0         | 400; 375; 350; 325      |
| 29,1         | 13,2         | 350; 325                |
| 29,1         | 13,1         | 375; 350; 325           |
| 29,1         | 13,0         | 375; 350; 325           |
| 29,1         | 12,9         | 375; 350                |
| 29,2         | 13,1         | 325; 300                |
| 29,2         | 13,0         | 375; 350; 325; 300      |
| 29,2         | 12,9         | 400; 375; 350; 325      |
| 29,2         | 12,8         | 375; 350                |
| 29,3         | 13,1         | 300                     |
| 29,3         | 13,0         | 325; 300; 275           |
| 29,3         | 12,8         | 375; 350; 325           |
| 29,4         | 13,0         | 275                     |
| 29,4         | 12,9         | 350; 300                |
| 29,4         | 12,8         | 300                     |

APPENDIX G

LEAST SQUARES FITS OF  $L_1f$  AND  $L_2f$  DATA - FROM FIGS. 5.7 AND 5.8

$L_1f$  vs  $Z_2$  (Using mean value of each column of data)

| $Z_2$          | $L_1f$           | $Z_2^2$          | $Z_2 \cdot L_1f$ |
|----------------|------------------|------------------|------------------|
| 275            | 205,4            | 75625            | 56485            |
| 300            | 203,8            | 90000            | 61140            |
| 325            | 202,9            | 105625           | 65943            |
| 350            | 202,2            | 122500           | 70770            |
| 375            | 200,6            | 140625           | 75225            |
| 400            | 199,2            | 160000           | 79680            |
| $\Sigma: 2025$ | $\Sigma: 1214,1$ | $\Sigma: 694375$ | $\Sigma: 409243$ |

(n = 6)

$$\Delta_0 = \begin{vmatrix} n & \Sigma Z_2 \\ \Sigma Z_2 & \Sigma Z_2^2 \end{vmatrix} = 65625$$

$$\Delta_1 = \begin{vmatrix} \Sigma Z_2 & -\Sigma L_1f \\ \Sigma Z_2^2 & -\Sigma Z_2 \cdot L_1f \end{vmatrix} = 14323613$$

$$\Delta_2 = \begin{vmatrix} n & -\Sigma L_1f \\ \Sigma Z_2 & -\Sigma Z_2 \cdot L_1f \end{vmatrix} = 3094,5$$

Now  $a/\Delta_1 = -b/\Delta_2 = 1/\Delta_0$

therefore  $a = -\Delta_1/\Delta_0 = 218,26$

$b = -\Delta_2/\Delta_0 = -0,0472$

thus 
$$L_1.f = a + bZ_2 = 218,26 - 0,0472 Z_2$$

$L_2f$  vs  $Z_2$

$Z_2$  : 275 300 325 350 375 400

$L_2f$  : 91,0 91,35 91,7 92,05 92,4 93,1

$$\sum Z_2 = 2025; \quad \sum L_2f = 551,6; \quad \sum Z_2^2 = 694375; \quad \sum Z_2.L_2f = 186340;$$

$$\Delta_0 = 65625; \quad \Delta_1 = 5678750; \quad \Delta_2 = 1050$$

therefore  $a = 86,5$  and  $b = 0,016$

thus 
$$L_2.f = 86,5 + 0,016 Z_2$$

APPENDIX H

CALCULATED IMPEDANCE DATA FOR A PARTICULAR ANTENNA CONFIGURATION  
(REFER TO FIG. 5.12)

$L_1 = 204,19\text{m}; \quad L_2 = 93,59\text{m}; \quad Z_2 = 350\Omega$

$k_0 h = \pi L_1 f / c = 2,1383 f_{\text{MHz}}$

| f<br>(MHz) | N  | $L_2/\lambda$ | $k_0 h$ | $Z_1$ (K-H)<br>( $\Omega$ ) | $Z_{1N} = Z_1/350$ |
|------------|----|---------------|---------|-----------------------------|--------------------|
| 1          | 1  | 0,312         | 2,1383  | 289+j755                    | 0,83+j2,16         |
| 2,01       | 2  | 0,627         | 4,298   | 91-j344                     | 0,26-j0,98         |
| 2,54       | 3  | 0,792         | 5,431   | 577+j932                    | 1,65+j2,66         |
| 3,52       | 4  | 1,098         | 7,527   | 110-j237                    | 0,31-j0,68         |
| 4,09       | 5  | 1,276         | 8,746   | 1110+j1049                  | 3,2+j3             |
| 5,03       | 6  | 1,569         | 10,76   | 122-j148                    | 0,35-j0,42         |
| 5,65       | 7  | 1,760         | 12,08   | 1860+j703                   | 5,31+j2            |
| 6,53       | 8  | 2,037         | 13,96   | 134-j93                     | 0,38-j0,27         |
| 7,20       | 9  | 2,256         | 15,39   | 2092-j219                   | 6-j0,63            |
| 8,04       | 10 | 2,508         | 17,19   | 151-j18                     | 0,43-j0,05         |
| 8,76       | 11 | 2,733         | 18,73   | 1555-j879                   | 4,44-j2,5          |
| 9,54       | 12 | 2,976         | 20,40   | 169+j37                     | 0,48+j0,11         |
| 10,32      | 13 | 3,220         | 22,07   | 973-j970                    | 2,78-j2,8          |
| 11,04      | 14 | 3,440         | 23,61   | 191+j90                     | 0,55+j0,26         |
| 11,88      | 15 | 3,706         | 25,40   | 620-j879                    | 1,77-j2,51         |

Smith Chart TA limits:

$\text{MAX} = S Z_4 / Z_2 = 2(50) / 350 = 0,286$

$\text{MIN} = Z_4 / S Z_2 = 50 / (2)(35) = 0,071$



## EXPERIMENTAL RESULTS FROM THE HORIZONTAL ANTENNA ABOVE GROUND

L1=26,51m (Phys) L2=12,2m (Phys) Z2=400 ohms at velocity factor  
0,897 Z4= 50 ohms

| Frequency<br>(MHz) | $Z_3$ ( $\Omega$ ) | $\theta_3$ (deg) | VSWR<br>(50 $\Omega$ ) |
|--------------------|--------------------|------------------|------------------------|
| 2,874              | 7,5                | -30              | 7,742                  |
| 3,170              | 68                 | -83              | 17,135                 |
| 3,398              | 22                 | -89              | 155,5                  |
| 3,500              | 39                 | -30              | 1,836                  |
| 3,534              | 34                 | 0                | 1,471                  |
| 3,650              | 47                 | 70               | 5,683                  |
| 3,800              | 35                 | 67               | 5,257                  |
| 4,263              | 255                | 72               | 17,080                 |
| 5,376              | 1500               | -28              | 33,985                 |
| 5,640              | 870                | -55              | 30,403                 |
| 6,288              | 290                | -65              | 14,061                 |
| 6,996              | 95                 | -40              | 2,812                  |
| 7,049              | 87                 | -34              | 2,370                  |
| 7,104              | 79                 | -31              | 2,107                  |
| 7,261              | 80                 | 0                | 1,600                  |
| 7,877              | 160                | 45               | 4,757                  |
| 8,783              | 255                | 29               | 5,885                  |
| 10,054             | 760                | 57               | 27,993                 |
| 10,097             | 760                | 58               | 28,773                 |
| 10,157             | 830                | 55               | 29,012                 |
| 10,753             | 2000               | 20               | 42,570                 |
| 11,030             | 2150               | -23              | 46,717                 |
| 12,448             | 430                | -69              | 24,281                 |
| 13,246             | 270                | -77              | 24,788                 |
| 14,000             | 91                 | -60              | 13,571                 |
| 14,198             | 51                 | -63              | 4,166                  |
| 14,348             | 45                 | -22              | 1,504                  |
| 14,458             | 53                 | 2                | 1,070                  |
| 14,521             | 75                 | 21               | 1,749                  |
| 15,306             | 190                | 33               | 4,629                  |
| 16,129             | 880                | 24               | 19,276                 |
| 17,064             | 410                | 58               | 15,640                 |
| 18,095             | 85                 | -35              | 2,372                  |
| 18,168             | 78                 | -28              | 1,990                  |
| 18,372             | 71                 | 1                | 1,421                  |
| 18,931             | 160                | 42               | 4,505                  |
| 20,388             | 420                | 34               | 10,178                 |
| 21,000             | 640                | 31               | 14,957                 |
| 21,259             | 760                | 28               | 17,232                 |
| 21,455             | 910                | 23               | 19,781                 |
| 21,504             | 930                | 20               | 19,800                 |
| 22,059             | 1225               | -21              | 26,249                 |
| 23,825             | 290                | -65              | 14,061                 |
| 24,835             | 120                | -54              | 4,573                  |
| 24,940             | 100                | -52              | 3,797                  |
| 25,365             | 52                 | 0                | 1,040                  |
| 26,883             | 730                | 34               | 17,637                 |
| 28,003             | 310                | -52              | 10,235                 |
| 28,500             | 620                | -55              | 21,713                 |
| 28,975             | 350                | -37              | 8,831                  |
| 29,383             | 280                | 1                | 5,601                  |
| 29,985             | 170                | 41               | 4,681                  |

### MEASUREMENT OF TRANSMISSION LINE CHARACTERISTIC IMPEDANCE

A 4,68 m length of nominally 300 $\Omega$  "tape" transmission line was measured using the configuration shown below.

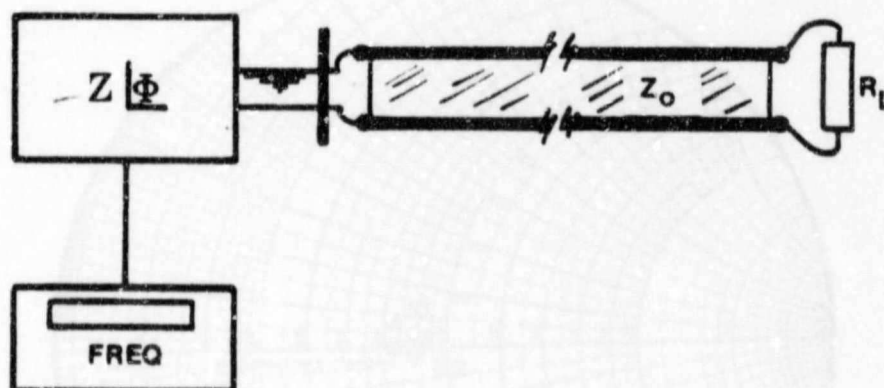


Figure J.1 : The measurement of cable characteristics

Two sets of measurements were made: the first with the equipment configured as above with a  $R_L = 300\Omega$  non-inductive resistor; the second set with a 1:1 Balanced-to-Unbalanced Transformer (BALUN) inserted between the probe and the transmission line (in view of the fact that the line is a balanced configuration while the Vector-Impedance meter  $Z|\Phi$  is unbalanced).

Case 1:  $R_L = 300\Omega$

| Frequency<br>(MHz) | $Z_{in}$<br>( $\Omega$ ) |
|--------------------|--------------------------|
| 2                  | $300 -4$                 |
| 5                  | $300 -10$                |
| 10                 | $250 5$                  |
| 20                 | $250 0$                  |
| 30                 | $240 -15$                |

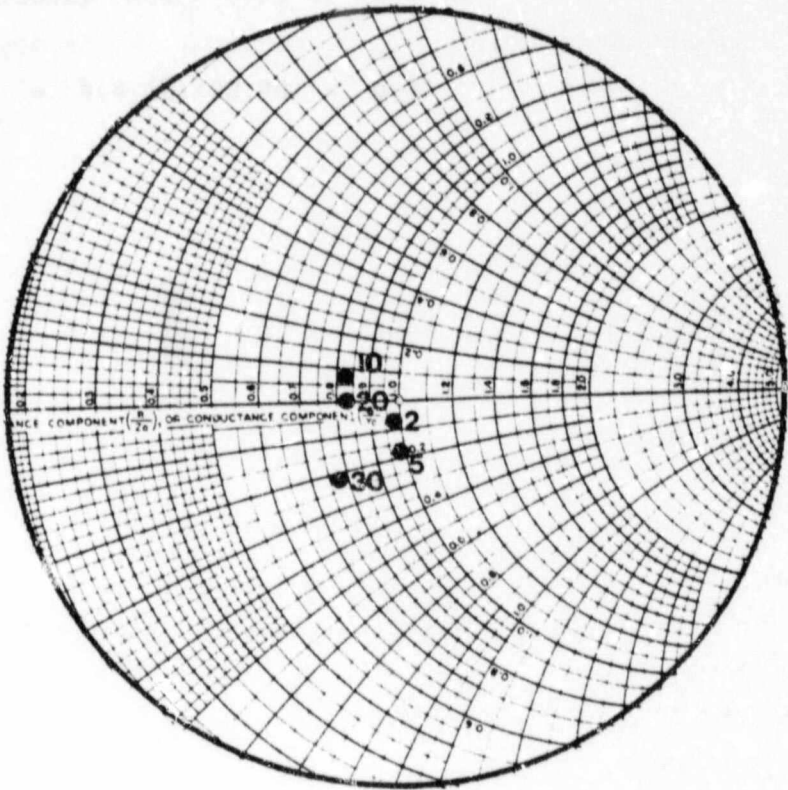


Figure J.2 : Data from the Table plotted on the Smith Chart normalized with respect to  $300\Omega$

Case 2: Quarterwave Transformer

Terminating the line with  $1800\Omega$  and 1:1 Balun between  $Z_0$  meter and line:-

$$\text{At } f = 13,056 \text{ MHz} \quad Z_{in} = 70 + j0\Omega$$

$$\text{therefore } L_{(elec)} = \lambda/4 = 4,68 \text{ m}$$

$$\text{hence } Z_0 = \sqrt{Z_{in} Z_L} = \sqrt{(70)(1800)} = 355\Omega$$

which confirms the above result in Case 1.

Now the freespace wavelength,  $\lambda_0$ , is

$$\lambda_0 = 300/13,056 = 22,98 \text{ m}$$

therefore Velocity Factor (VF) of line is

$$VF = \lambda/\lambda_0 = 4(4,68)/22,98 = 0,81$$

THE OPTIMISATION OF THE LINEAR DESIGN EQUATIONS FOR THE MULTIBAND  
WIRE ANTENNA

We commence with the equations obtained by means of the linear regressions in Appendix G:

$$L_1 f_{\min} = 218,26 - 0,0472 Z_2 \quad (K.1)$$

$$L_2 f_{\min} = 86,5 + 0,016 Z_2 \quad (K.2)$$

and use  $f_{\min} = 9,4$  MHz and  $L_1 = 22,0$  m with  $Z_2 = 355\Omega$ , from the experimental results in section 6.1.4.

$$\text{From (K.1)} \quad k_1 = (218,25 - (22)(9,4))/Z_2$$

where  $k_1$  is the corrected slope of the defining equation.

$$\text{Now } Z_2 = 355\Omega$$

$$\text{therefore } k_1 = 0,0323$$

Likewise (K.2), with  $L_2 = 9,95$  m, becomes,

$$L_2 f_{\min} = 86,5 + k_2(355)$$

$$\text{thus } k_2 = 0,0198$$

From which the three design equations are now given as,

$$L_1 f_{\min} = 218,26 - 0,0323 Z_2 \quad (K.3)$$

$$L_2 f_{\min} = 86,5 + 0,0198 Z_2 \quad (K.4)$$

and, after eliminating  $Z_2$ , and solving simultaneously,

$$L_1 f_{\min} = 359,19 - 1,631 L_2 f_{\min} \quad (K.5)$$

Likewise, instead of correcting the slopes of (5.4) and (5.5) we could have adjusted their points of intercept on their respective ordinates.

Thus from (5.4)

$$L_1 f_{\min} = A - 0,0472 Z_2$$

while from (5.6)

$$L_2 f_{\min} = B + 0,016 Z_2$$

Using  $Z_2 = 355\Omega$  we obtain

$$A = 223,56 \text{ and } B = 87,85$$

hence

$$L_1 f_{\min} = 223,56 - 0,0472 Z_2 \quad (K.6)$$

$$L_2 f_{\min} = 87,85 + 0,016 Z_2 \quad (K.7)$$

$$\text{and } L_1 f_{\min} = 482,72 - 2,95 L_2 f_{\min} \quad (K.8)$$

Now equating (K.3) and (K.6) we obtain

$$L_1 f_{\min} = 206,44 \quad (K.9)$$

$$\text{and } L_2 f_{\min} = 93,65 \quad (K.10)$$

which define  $L_1$  and  $L_2$  for given  $f_{\min}$  on the assumption that  $L_1$  and  $L_2$  can be chosen independently of  $Z_2$ . This is basically only true over a narrow range of  $Z_2$ . As Fig. 5.3 shows however, for  $350 \leq Z_2 \leq 400\Omega$  this assumption is reasonable.

The relationship between the frequencies on which the best match

occurs is obtained from the program "G5RV" with  $f_{\min} = 1$  MHz.  
 Thus from (K.9),  $L_1 = 206,44$  m and from (K.10),  $L_2 = 93,65$  m:

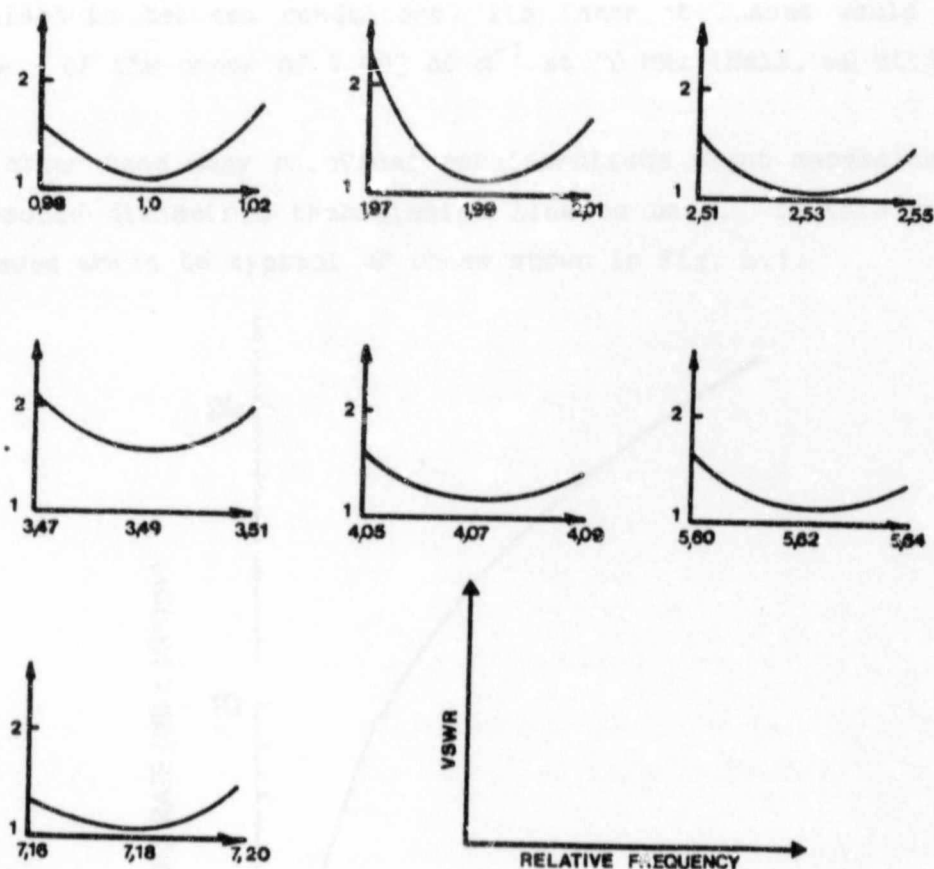


Figure K.1 : Predicted VSWR with frequency for the optimised antenna where  $L_1 = 206,44/f_{\min}$  and  $L_2 = 93,65/f_{\min}$

From Fig. K.1 the frequencies at which VSWR minima occur are related by

$$f_N = 1 : 1,99 : 2,53 : 3,49 : 4,07 : 5,62 : 7,18 : \dots$$



APPENDIX L

THE EFFECT OF TRANSMISSION LINE ( $L_2$ ) LOSS ON THE IMPEDANCE MATCHING MECHANISM

Any loss in the particular type of transmission line used for  $L_2$  will affect its performance as an impedance matching transformer. If  $L_2$  were constructed of "open-wire" line (i.e. with air dielectric between conductors) its inherent losses would be very low - of the order of  $0.003 \text{ dB m}^{-1}$  at 30 MHz (Hall, op cit).

On the other hand many practical considerations might necessitate that a solid dielectric transmission line be used. In this case line losses would be typical of those shown in Fig. L.1.

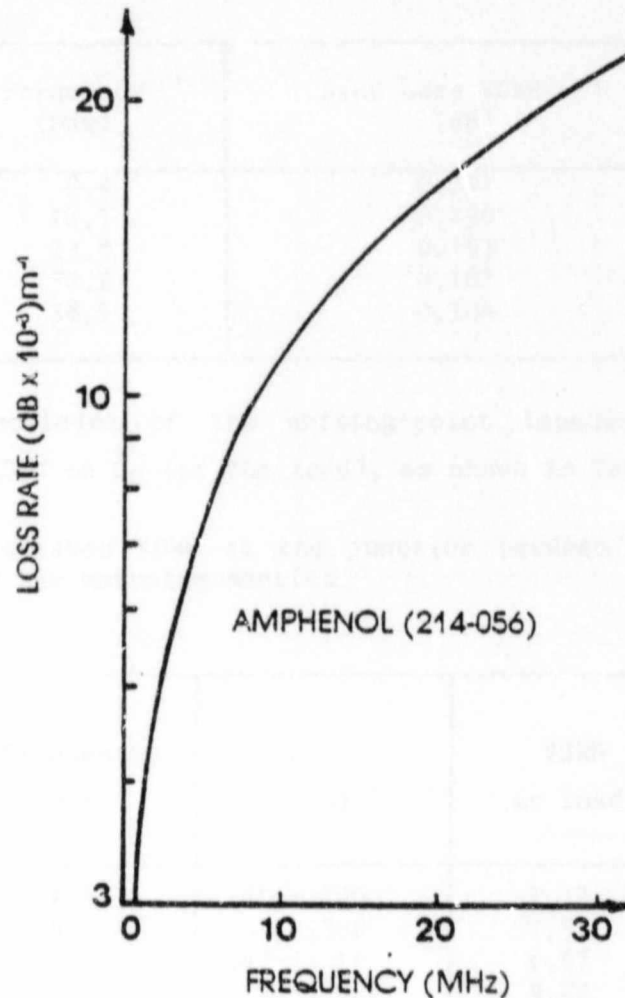


Figure L.1 : Typical loss with frequency for a solid-dielectric parallel conductor transmission line.

The fact that standing waves exist on  $L_2$ , because of its specific role as an impedance transformer, means that the intrinsic line losses given in Fig. L.1 will be increased because of the greater voltages and currents on the line compared to the matched case. These increased losses can be determined easily.

Considering the optimised antenna design (Fig. 6.14) as a working example, we proceed as follows:-

From Fig. L.1 the (typical) intrinsic losses, the frequencies on which the antenna is to operate are obtained and listed in Table L.1.

Table L.1 Intrinsic loss in  $L_2$  if matched

| Frequency<br>(MHz) | Line Loss VSWR = 1<br>(dB) |
|--------------------|----------------------------|
| 9,4                | 0,094                      |
| 18,7               | 0,136                      |
| 23,8               | 0,153                      |
| 32,9               | 0,187                      |
| 38,5               | 0,204                      |

Now from a knowledge of the driving-point impedance,  $Z_1$ , we calculate the VSWR on  $L_2$  (at the load), as shown in Table L.2.

Table L.2 Calculated VSWR at the junction between the antenna and the matching section

| Frequency<br>(MHz) | $Z_1$<br>( $\Omega$ ) | VSWR<br>at load |
|--------------------|-----------------------|-----------------|
| 9,4                | 310+j800              | 7,77            |
| 18,7               | 91-j336               | 7,59            |
| 23,8               | 667+j1016             | 6,67            |
| 32,9               | 110-j192              | 4,22            |
| 38,5               | 1589+j1065            | 6,57            |

The total loss,  $L_t$ , as a result of the standing waves are then obtained from (Hall, op cit, p.3-12):-

$$L_t = 10 \log \left| (B^2 - C^2)/B(1 - C^2) \right| \text{ dB}$$

where  $B = 10^{L_m/10}$

$L_m$  = line loss (dB) when matched (Table L.1)

$C = (S_l - 1)/(S_l + 1)$

$S_l$  = VSWR at load (Table L.2)

Table L.3    Calculated total loss on  $L_2$

| Frequency<br>(MHz) | $L_t$<br>(dB) |
|--------------------|---------------|
| 9,4                | 0,36          |
| 18,7               | 0,50          |
| 23,8               | 0,497         |
| 32,9               | 0,402         |
| 38,5               | 0,64          |

Note that within the HF band (3 to 30MHz) the loss on  $L_2$  never exceeds 0,5 dB.

By using the Smith Chart we show the effect of this loss on  $L_2$  and plot the point to which  $Z_{1N}$  at 18,7 MHz is transferred. This is shown in Fig. L.2, where the new value is  $Z'_3$ . It will be noted that this is still within the TA hence a suitable match will occur. We therefore conclude that line losses on  $L_2$ , under the conditions encountered in this antenna configuration can be ignored at HF.

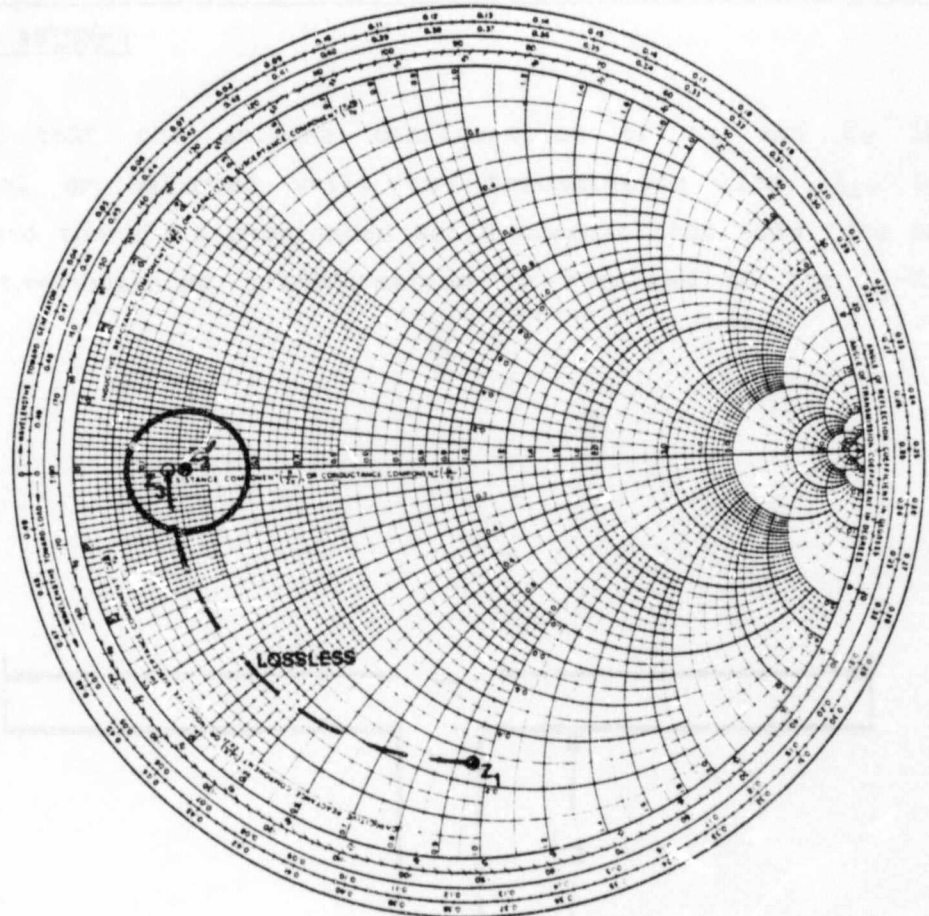


Figure L.2 : The effect of line losses at 18,7 MHz on the performance of the multiband antenna

THE USE OF A BALANCED-TO-UNBALANCED (BALUN) TRANSFORMER WITH THE MULTIBAND ANTENNA

The fact that the antenna configuration of  $L_1$  and  $L_2$  is symmetrical or balanced while the transmission line,  $Z_4$ , is coaxial and therefore unbalanced would suggest that some form of balanced-to-unbalanced transformation is required at the  $L_2$ - $Z_4$  junction.

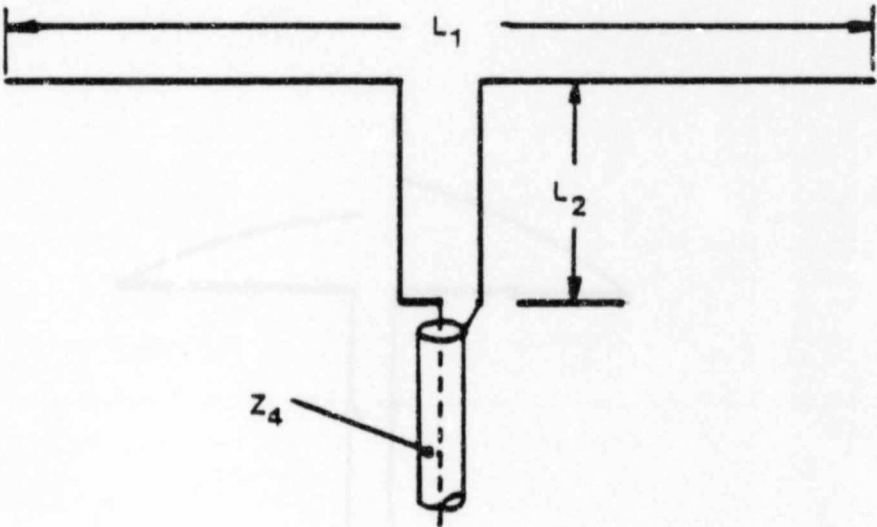


Figure M.1 : The junction between the balanced antenna system and the unbalanced transmission line.

The inherent symmetry means that equiamplitude, opposite-phase currents exist on both  $L_1$  and  $L_2$ , whereas at the junction with  $T_4$  a third current path, on the outside of the coaxial cable now exists. This is shown diagrammatically in Fig. M.2.

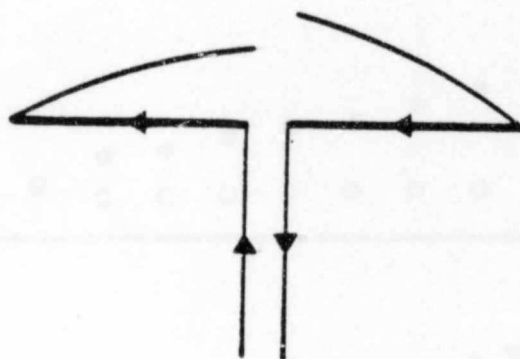
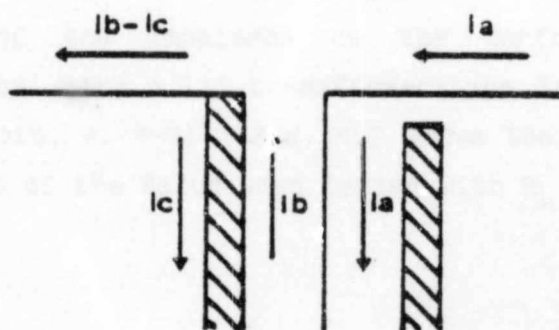


Figure M.2 : The unbalanced current distribution on the antenna caused by an unsymmetrical feed system (after Balanis, op cit, pp 216-217).

This lack of balance has two main consequences (Maxwell, 1983):-

- it causes distortion of the radiation pattern of the antenna;

it can affect the impedance (and hence the VSWR) presented to the terminal equipment;

The first of these is really of no consequence at HF with low directivity wire antennas close to both the ground and, in many cases, other undefined conducting objects likely to couple with the antenna and so alter its performance (Belrose, op cit; 1981).

The effect of the imbalance on the performance of  $L_2$  is investigated by using a 1:1 transformer-type Balun between  $Z_2$  and  $Z_4$  (Hall, op cit, p. 4-9). Fig. M.3 shows the measured impedance characteristic of the Balun when loaded with  $R_L = 50 \angle 0^\circ \Omega$ .

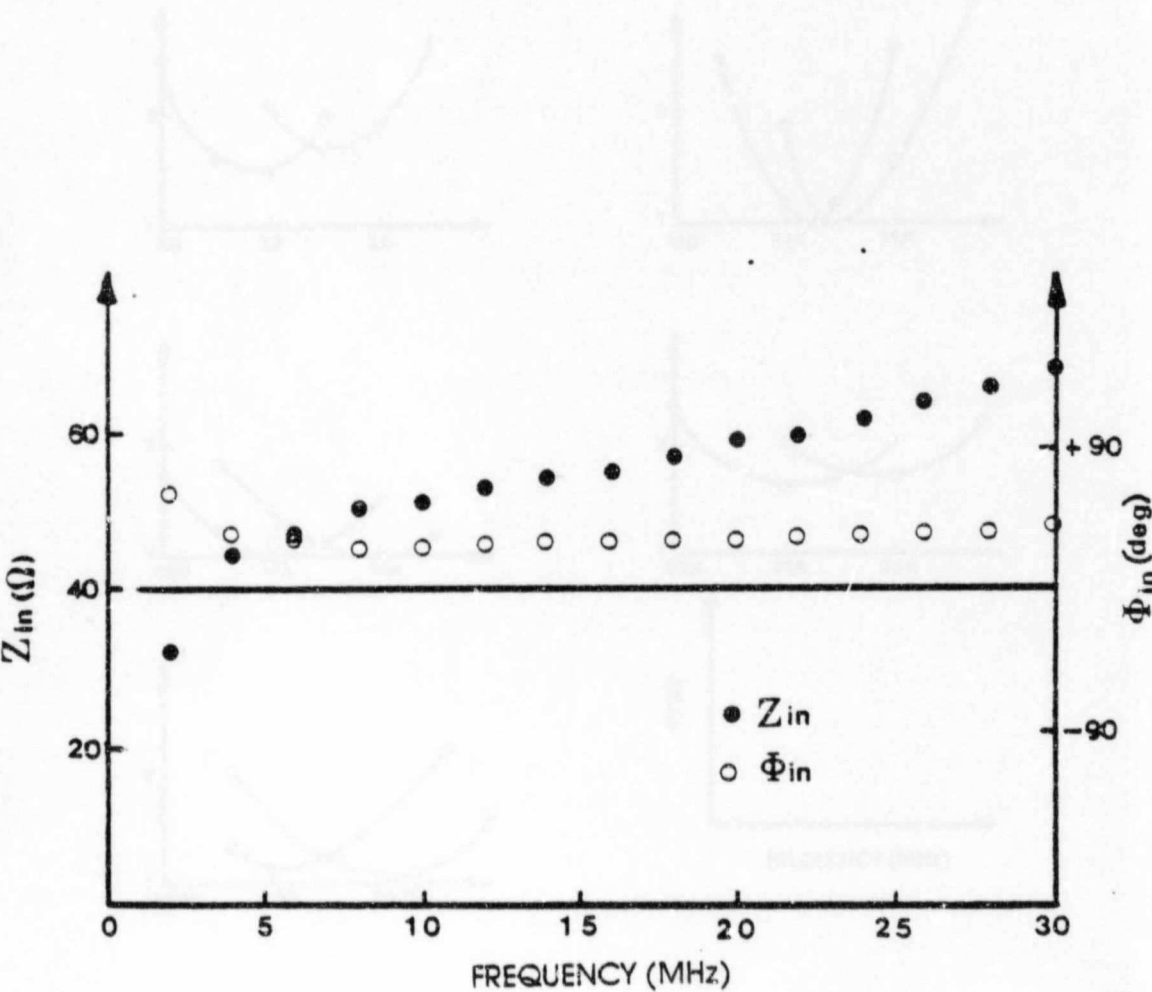


Figure M.3 : The measured performance of the 1:1 Balun when loaded with  $50\Omega$ .



These Balun characteristics are not ideal, in that the impedance transformation changes with frequency from about 0,8:1 at 3 MHz to nearly 1,4:1 at 30 MHz, while presenting about  $24\Omega$  of inductive reactance across the band. However, by its construction, the inherent balanced-to-unbalanced transformation characteristic, though not quoted nor measured, should be good.

The measured VSWR on five bands is shown in Fig. M.4 for the cases with and without the Balun.

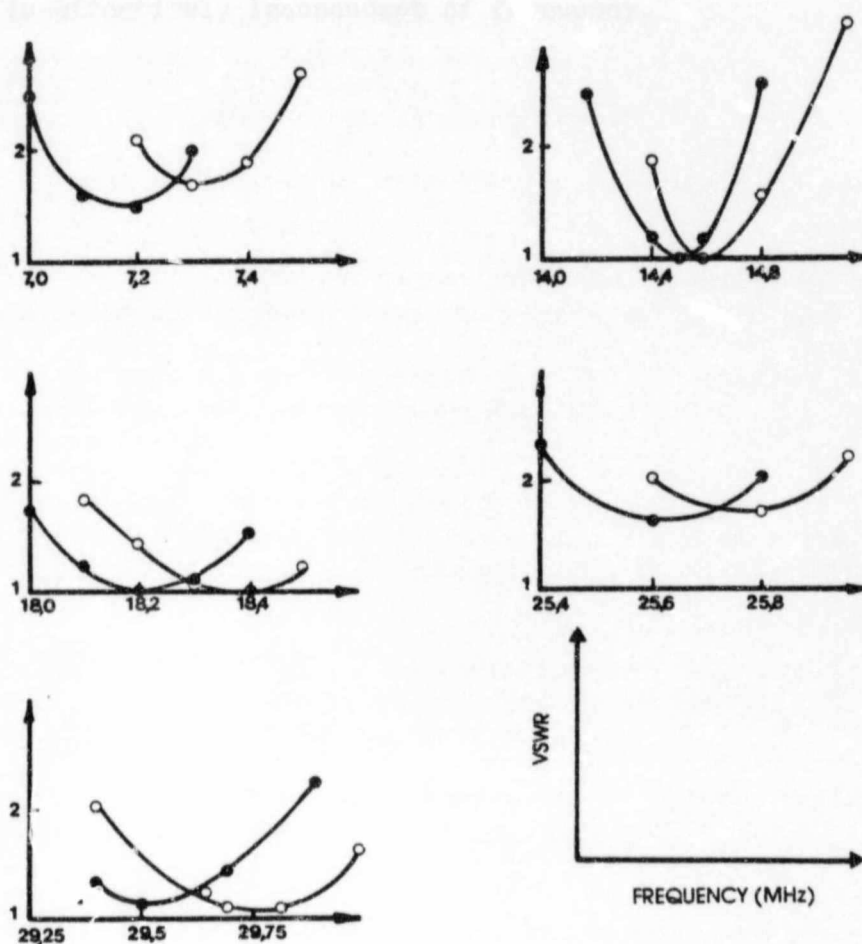


Figure M.4 : the comparative performance of the antenna system with and without a Balun.

- With Balun
- Without Balun

It will be noted from Fig. M.4 that the effect of the Balun is to lower the optimum frequency of the antenna system by about 1 per cent on each band. It is postulated that the decrease in frequency is probably due to the inductive loading provided by the Balun itself. In view of this a more acceptable technique may be to use a so-called Choke Balun (Maxwell, op cit, p. 40) which consists merely of a number of high permeability ferrite beads slipped on to the coaxial transmission line,  $Z_4$ , near its junction with  $L_2$ . This effectively increases the surface impedance of the braid thereby reducing the braid-outer current. This method, unlike other forms of choke devices (eg. the bazooka or sleeve balun) is effectively independent of frequency.

## REFERENCES:

1. Alanen, E. and Lindell, I.V. (1984) Impedance of vertical electric and magnetic dipole above a dissipative ground. **Radio Science**, vol. 19, no. 6, Nov.-Dec pp. 1469-1474.
2. Altshuler, E.E. (1961) The travelling-wave linear antenna. **IRE Transactions on Antennas and Propagation**, vol. AP-9, no.4, July, pp. 324-329.
3. Austin, B.A. (1983) Exercise Bushmen. **Certa Cito**, no. 3, Mar. pp. 6-10.
4. Austin, B.A. (1985a) Computer-aided design of a multiband dipole-based on the G5RV principle. **Radio Communication**, vol. 61, no. 8, Aug. pp. 614-624.
5. Austin, B.A. (1985b) Antennas and their environment - a systems approach. **Radio ZS**, vol. 39, no. 1, Jan. pp. 4-7.
6. Baker, D.C. (1985) Personal communication.
7. Balanis, C.A. (1982) **Antenna Theory** New York : Harper and Row, 1982.
8. Banos, A. (1966) **Dipole Radiation in the presence of a conducting half space**, Pergamon Press, New York. 1966.
9. Barclay, L.W. (1973) HF predictions and ionospheric forecasting. **Point-to-Point Communications**, vol. 17, no. 1, Jan. pp. 25-33.
10. Belrose, J.S. (1981) The effects of metal supporting towers on the radiation pattern of simple wire antennas. **IEE Conf. Publication (Antennas and Propagation)**, Pt 1, no. 195, pp. 84-87.
11. Belrose, J.S. ed (1983) The performance of antennas in their operational environment. **Agard Lecture Series No. 131**, NATO 1983, Specialized Printing Services, U.K.
12. Bennett, W.S. (1982) Characteristics of lossless resonant dipoles close to the earth. **Proc. IEEE**, vol. 70, no. 3, Mar. pp. 293-294.
13. Blake, L.V. (1984) **Antennas**, 2nd ed. Dedham : Artech House. 1984.
14. Braun, G. (1982) **Planning and engineering of shortwave links**, London: Heyden and Son. 1982.
15. Burke, G.J., Miller, E.K., Brittingham, J.N., Lager, D.L., Lytle, R.J. and Okado, J.T. (1981) Computer modelling of antennas near the ground. **Electromagnetics**, Jan. - Mar. vol. 1, no. 1, pp. 1-10.

## References (continued.....)

16. Burke, G.J. and Miller, E.K. (1984) Modelling antennas near to and penetrating a lossy interface. **IEEE Transactions on Antennas and Propagation**, vol. AP-32, no. 10, Oct. pp. 1040-1049.
17. Campbell, D.V. (1984) Personal computer applications of MININEC. **IEEE Antennas and Propagation Society Newsletter**, vol. 26, no. 1, Feb. pp. 5-9.
18. Carrel, R. (1961) The design of log periodic dipole antennas. **Antenna Lab. Technical Report**, 52, University of Illinois.
19. Chu, L.J. (1948) Physical Limitations of omnidirectional antennas. **Journal of Applied Physics**, vol. 19, no. 12, Dec. pp. 1163-1175.
20. Collins Radio Company (1959) **Fundamentals of Single Sideband**, 2nd ed. Collins Radio Co., Cedar Rapids, pp. 7.1-7.19.
21. Collins, B.S. (1980). The right antenna for the job. **Communications International**, vol. 7, no. 3, Apr., pp. 31-34.
22. Coutts, R.P. and Martin, A.L. (1981) High capacity, long haul, digital radio. **Journal of Electrical and Electronics Engineering, Australia**, vol. 1, no. 2, June, pp. 191-198.
23. Davies, K. (1965) **Ionospheric Radio Propagation**, National Bureau of Standards Monograph 80, US Government Printing Office, Washington.
24. Department of Communications (1984) General conditions governing the licensing and operation of radio communication stations and services (Australia), **DOC 68**, Canberra, pp. 1-21.
25. Dyson, J.D. (1962) Frequency-independent antennas. **Electronic**, vol. 35, no. 16, Apr. pp. 39-44.
26. Fitzgerald, R.G. (1965) Radiation efficiency of halfwave dipole antennas. **IEEE Transactions on Antennas and Propagation**, vol. AP-13, no. 2, Mar. pp. 326-327.
27. Gnab, V.K. (1966) Breitbandige rundstrahlantennen für kurzwellen. **Telefunken-Zeitung**, vol. 39, no. 2, Feb. pp. 186-193.
28. Gould, J.A. (1977) An automatic fast tuning hf matching unit for whip antennas. **Communication and Broadcasting**, vol. 4, no. 1, Spring, pp. 29-33.
29. Granger Assoc. (1967) Modern HF antennas - selection and application. **Granger Associates Technical Bulletin**, no. 1, 3rd ed. Palo Alto. 1967.

## References (continued.....)

30. Guertler, R.J.F. and Collyer, G.E. (1973) Improvements in travelling wave dipoles. **Proceedings IREE Convention**, Melbourne, Aug. pp. 70-71.
31. Hall, G.L. ed (1982) **The ARRL Antenna Book**, 14th ed. The American Radio Relay League, West Hartford, Con. 1982.
32. Hall, J. (1984) Maxcom antenna matcher and dipole cable kit. (Product Review). **QST**, vol. 68, no. 11, Nov. pp. 53-54.
33. Hansen, P.M. (1972) The radiation efficiency of a dipole antenna located above an imperfectly conducting ground. **IEEE Transactions on Antennas and Propagation**, vol. AP-20, no. 6, Nov. pp. 766-770.
34. Hansen, R.C. (1981) Fundamental limitations in antennas. **Proceedings IEEE**, vol. 69, no. 2, Feb. pp. 170-182.
34. Harrington, R.F. (1960) Effect of antenna size on gain, bandwidth and efficiency. **Journal of Res. Nat. Bur. Standards**, vol. 64D, Jan-Feb. pp. 1-12.
35. Harrington, R.F. (1967) Matrix methods for field problems. **Proceedings IEEE**, vol. 55, no. 2, Feb. pp. 136-149.
36. Harris, D.W. (1982) The Australian broadband dipole. **Amateur Radio**, vol. 40, no. 5, Apr. pp. 9-11.
37. Hartmann, P.R. (1981) Digital radio technology : present and future. **IEEE Communications Magazine**, vol. 19, no. 7, July 1981, pp. 10-14.
38. Hayward, W. (1976) Designing trap antennas. **QST** (Technical Correspondence), vol. 60, no. 9, Sept. pp. 38.
39. IEEE (1974) Standard Definitions of terms for Antennas. Std. 145 - 1973.
40. Johnson, W.C. (1950) **Transmission lines and networks**. New York, McGraw-Hill, 1950.
41. Jones, J.E. (1976) Analysis of the symmetric centre-fed V-dipole antenna. **IEEE Transactions on Antennas and Propagation**, vol. AP-24, no. 3, May, pp. 316-322.
42. Jordan, E.C. and Balmain, K.G. (1968) **Electromagnetic Waves and Radiating Systems**, 2nd ed. New Jersey: Prentice-Hall Inc. 1968.
43. Jordan, E.C., Deschamps, G.A., Dyson, J.D. and Mayes, R.E. (1964) Developments in broadband antennas. **IEEE Spectrum**, vol. 1, no. 4, Apr. pp. 58-71.



## References (continued....)

44. Julian, A.J., Logan, J.C. and Rockway, J.W. (1982) **MININEC : A mini-numerical electromagnetics code**. Technical Document 516, Naval Ocean Systems Centre, San Diego, 1982.
45. Kandoian, A.G. (1968) ed. **Reference data for radio engineers**, 5th ed. Indianapolis Howard W. Sams. 1968.
46. King, R.W.P. (1967) The linear antenna - eighty years of progress. **Proceedings IEEE**, vol. 55, no. 1, Jan. pp. 2-16.
47. King, R.W.P. and Harrison, C.W. (1969) **Antennas and Waves - A modern approach**. Cambridge: The MIT Press, 1969.
48. King, R.W.P. (1983) The wave antenna for transmission and reception. **IEEE Transactions on Antennas and Propagation**, vol. AP-31, no. 6, Nov. pp. 956-965.
49. Krauss, H.L., Bostian, C.W. and Raab, F.H. (1980) **Solid State Radio Engineering**. New York: John Wiley and Sons. 1980.
50. Kraus, J.D. and Carver, K.R. (1981) **Electromagnetics** 2nd ed. New York: McGraw-Hill, 1981.
51. Lee, K.S.H., Marin, L. and Costillo, J.P. (1976) Limitations on wire-grid modelling of a closed surface. **IEEE Transactions on Electromagnetic Compatibility**, vol. EMC-18, no. 3, Aug. pp. 123-129.
52. Lewis and Underhill (1979) Two stage tuning and matching of HF mobile antennas. **Proceedings IERE Conference on Land Mobile Radio**, Lancaster, U.K. Sept. pp. 213-221.
53. Lomer, G.J. (1978) Combat net radio for the 1980s. **Journal Royal Signals Institution**, vol. 13, no. 5, Summer, pp. 4-7.
54. MacLean, T.S.M. (1968) Impedance properties of capacitively loaded dipoles, **Proceedings IEE**, vol. 115, no. 10, Oct. pp. 1411-1418.
55. Maxwell, W. (1983) Some aspects of the Balun problem. **QST**, vol. 67, no. 3, Mar. pp. 38-40.
56. Moore, J. and Pizer, R. (1984) **Moment methods in electromagnetics**, Letchworth: Research Studies Press, J. Wiley and Sons, 1984.
57. Morgan, H.K. (1940) A multifrequency tuned antenna system. **Electronics**, vol. 13, Aug. pp. 42-48.
58. Nicholson, T. (1981) Compact multiband antenna without traps. **QST**, vol. 65, no. 11, Nov. pp. 26-27.
59. Nicol, J.L. (1980) The input impedance of horizontal antennas above an imperfect plane earth. **Radio Science**, vol. 15, no. 3, May-June, pp. 471-477.

## References (continued.....)

60. Noonan, C.R.M. (1981) H.F. communications in defence, **Communications and Broadcasting**, vol. 8, June, pp. 15-20.
61. Nyquist, D.P. and Chen, K. (1968) The travelling-wave linear antenna with non-dissipative loading. **IEEE Transactions on Antennas and Propagation**, vol. AP-16, no. 1, Jan. pp. 21-31.
62. NITR (National Institute for Telecommunications Research, CSIR), **Radio Propagation predictions for Southern Africa**, Johannesburg, CSIR, Series ETP.
63. Ormondroyd, R.F. (1978) Spread-spectrum communication in the land mobile service. **Proceedings COM.78 Conference**, Birmingham, U.K., Publ. No. 162, pp. 156-162.
64. Popovic, B.D., Dragovic, M.B. and Djordjeric, A.R. (1982) **Analysis and synthesis of wire antennas**, Chichester: Research Studies Press, J. Wiley and sons, 1982. pp. 109-144.
65. Parhami, P. and Mittra, R. (1980) Wire antennas over a lossy half-space. **IEEE Transactions on Antennas and Propagation**, vol. AP-28, no. 3, May, pp. 397-403.
66. Raggett, R.J. ed. (1983) **Jane's Military Communications**, 4th ed., London: Jane's Publishing Company Ltd. 1983
67. Rahmet-Samii, Y., Mittra, R. and Parhami, P. (1981) Evaluation of Sommerfeld integrals for lossy half-space problems. **Electromagnetics**, vol. 1, no. 1, Jan-Mar. pp. 1-28.
68. Ramsdale, P.A. (1977) Signal/noise-ratio performance of loaded wire antennas. **Proceedings IEE**, vol. 124, no. 10, Oct. pp. 840-844.
69. Ramsdale, P.A. (1978) Wire Antennas, in **Advances in Electronics and Electron Physics**, Academic Press, Inc. 1978, pp. 123-196.
70. Ramsdale, P.A. (1979) Antennas - an overview. **Communications International**, vol. 6, no. 10, Oct. pp. 22-28.
71. Rao, B.L.J., Ferris, J.E. and Zimmerman, W.E. (1969) Broadband characteristics of cylindrical antennas with exponentially tapered capacitive loading. **IEEE Transactions on Antennas and Propagation**, vol. AP-17, no. 2, Mar. pp. 145-151.
72. Ribchester, E. (1978) Trends in tactical communications. **Electron (G.B.)**, vol. 143, 21 Aug. pp. 19-21.
73. Richmond, J.H. (1966) A wire-grid model for scattering by conducting bodies. **IEEE Transactions on Antennas and Propagation**, vol. AP-14, no. 6, Nov. pp. 782-786.



## References (continued.....)

74. Rudge, A.W. (1972) Input impedance of a dipole antenna above a conducting half space. **IEEE Transactions on Antennas and Propagation**, vol. AP-20, no. 1, Jan. pp. 86-89.
75. Rumsey, V.H. (1966) **Frequency independent antennas**, New York: Academic Press, 1966.
76. Sanne, M.C. and Marock, M. (1984) **Broadband wire antennas**, unpublished BSc (Eng) dissertation, University of the Witwatersrand, Johannesburg, 1984.
77. Scholtz, R.A. (1977) The Spread Spectrum concept. **IEEE Transactions on Communications**, vol. COM-25, no. 8, Aug. pp. 748-755.
78. Sinclair, G. (1948) Theory of models of electromagnetic systems. **Proceedings IRE**, vol. 36, no. 11, Nov. pp. 1364-1370.
79. Smith, D.L. (1975) Trap-loaded cylindrical antenna. **IEEE Transactions on Antennas and Propagation**, vol. AP-23, no. 1, Jan. pp. 20-27.
80. Smith, G.S. (1977a) Efficiency of Electrically small antennas combined with matching networks. **IEEE Transactions on Antennas and Propagation**, vol. AP-25, no. 3, May, pp. 369-373.
81. Smith G.S. (1977b) An analysis of the Wheeler method for measuring the radiating efficiency of antennas. **IEEE Transactions on Antennas and Propagation**, vol. AP-25, no. 4, July, pp. 552-556.
82. Stutzman, W.L. and Thiele, G.A. (1981) **Antenna Theory and Design**, New York: John Wiley and Sons, 1981.
83. Sweeting, M.N. and Davis, Q.V. (1980) Electrically short HF aerial systems. **Agard Conference Proceedings**, no. 263, pp. 48.1-48.21.
84. TCI (1980) Rooftop-rotatable log-periodic antennas - model 628, **TCI advertising literature**, 4/80, California, 1980.
85. Terman, F.E. (1943) **Radio Engineers Handbook**, New York: McGraw-Hill, 1943 pp. 854.
86. Theron, J.C. (1983) Tactical antennas and their application in the military environment. **Proceedings SAIEE Symposium on Antennas and Propagation**, 16-18 May, pp. V.1-V.9.
87. Treharne, R.F. (1983a) Multipurpose whole-band HF antenna architecture. **Journal of Electrical and Electronics Engineering, Australia**, vol. 3, no. 2, June, pp. 141-153.

## References (continued.....)

88. Treharne, R.F. (1983b) Natural wave logarithmically truncated Beverage antenna. **ibid**, vol. 3, no. 3, Sept. pp. 223-227.
89. Underhill, M.J. (1981) Wide range antenna matching networks. **Proceedings IERE Conference on Radio Receivers and Associated Systems**, 7-9 July, pp. 101-135.
90. Varney, L. (1958) An effective multiband aerial of simple Construction. **RSGB Bulletin**, vol. 34, no. 7, July, pp. 19-20.
91. Vice, R.W. (1954) A survey of ground wave propagation conditions in South Africa. **Transactions SAIEE**, vol. 45, no. 4, Apr. pp. 139-159.
92. Vogler, L.E. and Noble, J.L. (1964) Curves of input impedance change due to ground for dipole antennas. **National Bureau of Standards Monograph 72**, Jan. pp. 1-47.
93. Wait, J.R. (1969) Characteristics of antennas over lossy earth, in **Antenna Theory**, Pt II, edited by Collin R.E. and Zucker, F.J., New York: McGraw-Hill, 1969 pp. 386-437.
94. Wait, J.R. (1976) Long-wave behaviour of the Beverage wave aerial. **Electronics Letters**, vol. 12, no. 14, July, pp. 358-359.
95. Watterson, C.C. (1979) Methods of improving the performance of HF digital radio systems. **NTIA Report 79-29**, U.S. Department of Commerce, Oct.
96. Wheeler, H.A. (1959) The radiansphere around a small antenna. **Proceedings IRE**, vol. 47, no. 8, Aug. pp. 1325-1331.
97. Wu, T.T. (1961) Theory of the dipole antenna and two-wire transmission line. **Journal Mathematical Physics**, vol. 2, no. 4, July-Aug., pp. 550-574.
98. Wu, T.T. and King, R.W.P. (1965) The cylindrical antenna with non-reflecting resistive loading. **IEEE Transactions on Antennas and Propagation**, vol. AP-13, no. 3, May, pp. 369-373.



**Author** Austin Brian Anthony

**Name of thesis** The Analysis, Design And Evaluation Of A Multiband Wire Antena For Tactical And Point-to-point High Frequency Communications With Special Reference To The Role Of The Ionosphere An. 1985

***PUBLISHER:***

University of the Witwatersrand, Johannesburg

©2013

***LEGAL NOTICES:***

**Copyright Notice:** All materials on the University of the Witwatersrand, Johannesburg Library website are protected by South African copyright law and may not be distributed, transmitted, displayed, or otherwise published in any format, without the prior written permission of the copyright owner.

**Disclaimer and Terms of Use:** Provided that you maintain all copyright and other notices contained therein, you may download material (one machine readable copy and one print copy per page) for your personal and/or educational non-commercial use only.

The University of the Witwatersrand, Johannesburg, is not responsible for any errors or omissions and excludes any and all liability for any errors in or omissions from the information on the Library website.